

REMARKS

Claim 1 was rejected for lack of particularity. Claims 1 and 6 were rejected as unpatentable over Smith in view of Marko. Claims 3 and 4 were rejected as unpatentable over Smith in view of Marko in further view of Carlson. Claim 5 was rejected as unpatentable over Smith in view of Marko in further view of Kim. Claim 7 was rejected as unpatentable over Smith in view of Marko in further view of Rattlingourd. Claim 8 was rejected as unpatentable over Smith in view of Marko in further view of Rattlingourd in further view of Kim in further view of Tucci. Applicant requests reconsideration.

It is not understood what is "page 3, paragraph [0023] lines 3-12. The specification was filed with page numbers and line numbers, and those should be used. The "adjusted timing signals" are incorrectly stated, and should be "adjusted timing pulses" (14). (See Figure 1 and claim 1) While the entire timing recovery loop is used to produce the adjusted timing pulses, the baseband signal 10 encodes the original bit timing. The specification teaches " The binary data and timing are encoded into the baseband signal waveform 10. The baseband waveform 10 is fed into a conventional data detector 11 and a pulse detector and comparator 12. The pulse detector and comparator 12 generates adjusted timing pulses 14 that can be used by the data detector 11 to sample the baseband waveform 10 to reconstruct the digital bit stream 15." It is the transition pulses detected from the baseband signal that give rise to the adjusted timing pulses. Hence, the specification teaches that the baseband signal includes the timing information

1 from which the adjusted timing pulses are generated. In Figure 1,
2 there is a path extending from the baseband waveform 10 through the
3 pulse detector and comparator 12 that generates the adjusted timing
4 pulses 14. The specification and claims reflect this necessary
5 path. Claim 1 is particular and accurate in this regard.

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7 The timing pulse delay adjustor 16 does adjust the adjusted
8 timing pulse delay routed from the adjuster 26 to the pulse and
9 comparator detector 12. Claim 1 does not refer to a reference
10 signal, as intended, nor is a reference signal found in Figure 1.
11 Claim 1 vis-à-vis Figure 1 is accurate. Hence, the adjustor 26
12 adjusts the adjusted timing pulse delay, as claimed in claim 1, as
13 described in the specification, and as shown in Figure 1. Claim 1
14 reads directly on Figure 1, without regard to an embedded reference
15 clock signal in the pulse detector and comparator 12. However,
16 Claim 1 was amended to more clearly recite that the timing pulse
17 delay adjustor 26 adjusts the adjusted timing pulse delay.

18

19 Applicant extends appreciation for the thorough examination.
20 The cited references describe improvements in early-and-late gate
21 timing recovery. There are major differences between the early-and-
22 late gate approach of the cited references and the claimed random
23 walk filter. The early-and-late gate approach essentially detects
24 the times of ascending crossing and descending crossing of a
25 threshold level. This is well-known threshold detection. By noting
26 the time differences between the estimated time at the center of
27 the pulse and the early and late times of threshold crossing, the
28 estimated time of pulse center is adjusted. The random walk filter,

1 on the other hand, counts the number of times over many data
2 periods that the baseband waveform pulses that are obtained by
3 differentiating the baseband waveform, lead or lag the
4 corresponding delayed or adjusted reference timing pulses.

5

6 The random walk filter does not determine the amount of lead
7 or lag for each pair of pulses that could be had by simple
8 threshold detection, as in the cited references. The random walk
9 filter sums, over many data periods, accumulative +1's, as a the
10 total count for leads of any amount and -1's for lags of any
11 amount, for indicating whether any adjustment to the reference
12 timing pulse delay is needed, for improved synchronization with the
13 baseband waveform. When the delayed reference timing pulses are
14 synchronized with the baseband waveform, +1, -1 and 0 would be
15 randomly distributed over the many data periods, with the total
16 count resembling the Brownian motion, thus resulting in the name of
17 random walk filter approach to timing recovery. This is the major
18 difference between early-and-late gating and random walk filtering.

19

20 The random walk filter has a timing window that initially
21 covers an entire data period, so that the random walk filter can
22 always find the baseband waveform pulse corresponding to a
23 particular adjusted reference timing pulse. The window can decrease
24 to very small size as the baseband waveform is acquired and
25 tracked. By contradistinction, the early-and-late gate window is
26 centered at the pulse center and will always register each and
27 every early and late crossings of the threshold. Depending on the

1 threshold level relative to the peak value, the window cannot be
2 decreased below a specific size.

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4 Thresholding has different meanings between the early and late
5 gate and the random walk filter. The threshold for the early-and-
6 late gate determines the time, relative to the pulse center, for
7 the crossings. By contradistinction, the random walk filter uses
8 two thresholds, the first threshold allows the count to be reset
9 and the second threshold is set so that any count below that
10 threshold would not result in any adjustment to the delay. This
11 ensures that the random walk filter would not make adjustments
12 based on noise. Thus, threshold for the early-and-late gate
13 approach is an amplitude detection threshold while the two
14 thresholds for the random walk filter approach are counting
15 thresholds.

16

17 While the random walk filter may include conventional
18 components as in the cited references, the random walk filter
19 provides a fundamentally different approach to timing recovery. The
20 random walk filter implements accumulative lead and lag counting
21 over many data bits to achieve timing recovery, whereas the cited
22 references use conventional localized threshold detection for
23 timing about a pulse center. The cited references do not teach nor
24 suggest accumulative lead and lag counting using random walk
25 filtering over a large time period.

26

27 It is true that the reference timing pulses are locally
28 generated. However, the adjusted timing pulses are corrected by the

1 random walk filter so that the adjusted timing pulses are
2 synchronized with the baseband waveform for accurate bit detection.
3

4 The timing pulse delay adjuster (26) does not delay the adjusted
5 timing pulses. It delays the reference timing pulses to produce the
6 adjusted timing pulses to be synchronized with the baseband
7 waveform.

8
9 Smith proposed a technique based on early-and-late gate phase
10 comparator and reversible binary counter.

11
12 Marko also proposed an early-and-late gate technique using two
13 digital phase lock loops with narrowband and wideband filters.
14 These versions of the early-and-late gating are different to the
15 random walk filter approach in several aspects.

16
17 Carlson also used the early-and-late gate technique by setting
18 the early and late windows and detecting the timing data relative
19 to the window centers. Again, while using some of the same
20 components, this early-and-late gate approach is quite different
21 than the random walk filter approach.

22
23 Kim used a training sequence with a predetermined pattern to
24 synchronize incoming waveform with clock periods. Lead or lag
25 information is then used for clock adjustment. This is a technique
26 for acquisition before data waveform is received while the random
27 walk filter approach performs timing recovery while processing data
28 waveform.

1 Rattlingourd used an Oscillator to generate a signal at a fixed
2 frequency (f), which is much larger than the data rate. Through a
3 Variable Divider with value m (e.g., 200), a clock output signal is
4 produced with rate f/m that is close to the data rate. A Phase
5 Detector is then used to detect differences between the positive
6 going and negative going edges of the data and the clocking edges
7 of the clock signal. Its output drives an Up/Down Counter. A Window
8 Counter also compares data edges with the clocking edges to assess
9 the scope of the timing error. The outputs of the Up/Down Counter
10 and Window Counter modify the value m of the Variable Divider. If
11 there is no difference between data edges and clocking edges, the
12 value m remains the same. When there is a difference, m would
13 increase or decrease. Increasing m by unity essentially delays the
14 arrival of the next clocking edge by one oscillator cycle, and the
15 next clocking edge after that by 2 oscillator cycles, and so forth.
16 Decreasing m has the opposite effect. The clock output signal rate
17 is also affected by a change in the value m . The random walk filter
18 approach, on the other hand, provides a time delay that shifts all
19 the timing pulses. The delay is adjusted through by the output of
20 the random walk filter. Thus, there are four differences between
21 the approaches. Rattlingourd's technique adjusts the quantity f/m ,
22 which is the estimated data rate, while the random walk filter
23 adjusts delay. Rattlingourd's oscillator generates a frequency much
24 higher than the data rate, while random walk filter generates
25 timing pulses at the data rate. Rattlingourd's approach requires
26 that the timing comparison between data edges and clocking edges
27 occurs at the clocking edges of the clock signal. The random walk
28 filter allows the counting of a data transition pulse within a

1 large window, of the order of a data period, centered at the
2 corresponding timing pulse. Rattlingourd's approach requires the
3 adjustment of the parameter m to occur at the edges of the clock
4 signal, while in random walk filter, the delay can be adjusted at
5 any time. Rattingourd describes leads and lag with up and down
6 counting and adjusts recovery timing by changing pulse separation,
7 whereas, the random walk filter recovery loop adjust the timing
8 pulse delay delaying the data stream with equal and fixed bit
9 separation. Rattingourd does not suggest and teaches contrary to
10 equal and fixed bit separation using an adjusted timing pulse
11 delay.

12

13 Tucci used a bank of delay lines to capture the lead and lag
14 signals, a phase locked loop to track the timing error, and a phase
15 comparator to make adjustment. This approach is much like the
16 early-and-late gate technique in that it estimates the amount of
17 lead and lag, and adjusts accordingly. Therefore, while using some
18 of the same components, this approach is quite different than the
19 random walk filter approach.

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The cited references teach early-and-late gating using threshold detection over a data period for timing recovery. The random walk filter using clock counting over many time periods to shift the adjusted timing pulses. The invention proceeds contrary to the cited references and cited reference are thus strong evidence of non-obviousness. Allowance of the claims is requested.

Respectfully Submitted

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